

INTRODUCTION

DESCRIPTION

The IN-PLUG® IPS25 Integrated Circuit is a self-powered precision feedback controller for fly-back off-line switching power supply applications. It can be powered from the load voltage even under short-circuit conditions, therefore saving the auxiliary circuitry normally required to power it. Especially designed to address very low-cost, low-power fly-back power supplies which output current does not exceed approximately 800 mA. It can be used for:

- Voltage regulation / current limiting
or
- Current regulation / voltage limiting.

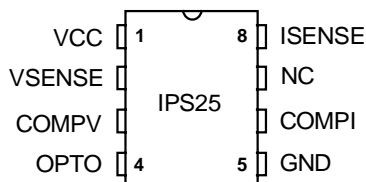
It contains the following features:

- Trimmed band-gap references.
- Voltage & current error amplifiers.
- Loop compensation circuitry for both V & I.
- Opto drivers.

This low-cost IC has been optimized to operate in conjunction with the IN-PLUG® IPS10/IPS15 primary SMPS off-line driver ICs, for domestic and international inputs ranging from 85V to 264V AC.

The IPS25 feedback controller can also be used for different applications where the primary circuitry is built around discrete components or standard ICs other than AAI's IPS10/IPS15 ICs.

PIN CONFIGURATION: DIP-8 / SOIC-8



FEATURES

- Lowest cost solution for low-power off-line flyback applications.
- High performance SMPS.
- Full flexibility for loop stability control.
- Operates with optocoupler for maximum isolation from line voltage.

APPLICATIONS

- Standby power supplies for TV, VCR and IR remotely-controlled appliances.
- Cordless and feature phones.
- Cellular phone chargers.
- Power tools fast chargers with trickle and on/off.
- Laptops and personal digital assistants.
- Utility meters.
- Cost-effective replacements for bulky plug-in transformers.

ORDERING INFORMATION

Part No.	ROHS / Pb-Free	Package	Temperature Range	
IPS25C-D	-G-LF	8-Pin PDIP	0°C to +70°C	Commercial
IPS25I-D	-G-LF	8-Pin PDIP	-40°C to +85°C	Industrial
IPS25C-SO	-G-LF	8-Pin SOIC	0°C to +70°C	Commercial
IPS25I-SO	-G-LF	8-Pin SOIC	-40°C to +85°C	Industrial

For detailed ordering information, see page 12

PIN DESCRIPTION

Number	Name	Description
1	VCC	IC positive supply pin. The IC is powered from this pin which is connected to the positive output of the SMPS and powers the chip during both normal operation and short-circuit conditions.
2	VSENSE	Voltage sensing pin. Inverting input of the voltage error amplifier which positive input is connected to an internal trimmed 2.57V voltage reference. An external voltage divider connected to this point sets the output voltage. This pin is also used for voltage loop compensation when required.
3	COMPV	Voltage loop compensation pin. This is the output of the voltage error amplifier. The loop compensation network, when required, is connected between this point and VSENSE pin. Please note that this pin is not a zero-impedance node.
4	OPTO	Optocoupler driver pin. This pin drives an external optocoupler connected to GND. A current-mode drive is used for maximum noise rejection.
5	GND	Ground pin. This is the most negative IC pin. The first output decoupling capacitor should return to it. Do not confuse this pin with the ISENSE pin which is connected to the negative output of the SMPS.
6	COMPI	Current loop compensation pin. This pin is used to compensate the current-limiting loop. In order to prevent conflicts with the voltage loop, this compensation is performed on an intermediate stage which impedance is approximately 400K Ω .
7	NC	Not connected pin (spare). This pin is not internally connected in the IPS25. It is available to provide additional features when ordering custom variations of the basic IPS25.
8	ISENSE	Current sensing pin. This pin is connected to the negative output of the SMPS. A current-sensing resistor connected between this point and GND is used to sense the output current. Current limiting occurs when the voltage developed reaches 1.33V. Please note that the second output decoupling capacitor should return to this pin, not to GND.

IN-PLUG® IPS25 FUNCTIONAL DESCRIPTION

The IN-PLUG® IPS25 Integrated Circuit was especially designed for voltage regulation / current limiting or current regulation / voltage limiting in very low-cost, low-power fly-back power supply applications which output current does not exceed approximately 800 mA.

It is a self-powered precision feedback controller for fly-back off-line switching power supply applications. It can be powered from the load voltage even under short-circuit conditions, therefore saving the auxiliary circuitry normally required to power it.

It provides precise and temperature-stable voltage and current regulations that can be fully and independently adjusted.

It sources the current necessary to drive a ground-referred optocoupler for maximum noise rejection and low output ripple.

The principal features are:

- Trimmed band-gap references.
- Voltage & current error amplifiers.
- Loop compensation circuitry for both V & I.
- Opto drivers.

Secondary circuitry:

(please refer to the application schematic on page 2)

- 1) The chip is powered through the forward operation of the flyback transformer which ensures proper operation on a dead short-circuit on the output.
- 2) The current limiting is adjustable through R8
- 3) The regulated voltage is adjustable through R6/R7
- 4) COMPI & COMPV pins are available for loop compensations. Current loop compensation is always required whereas voltage loop compensation can usually be omitted when operating in conjunction with the IN-PLUG® IPS10/IPS15 flyback controller ICs.

Application informations:

▪ Determination of V_{out} (Output Voltage):

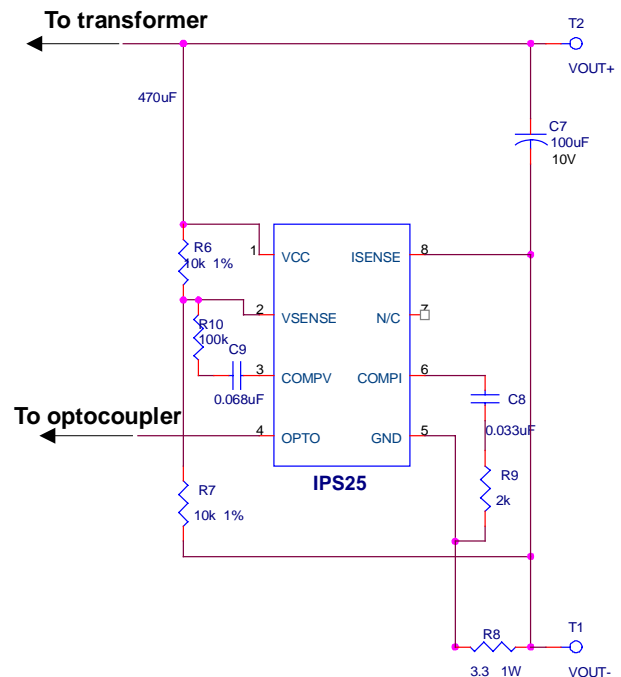
$$V_{out} = V_{SENSE} \times (R6+R7) / R7$$

(Typical $V_{SENSE} = 2.60V$)

▪ Determination of I_{limit} (Current Limit):

$$I_{limit} = I_{SENSE} \text{ threshold} / R8$$

(Typical $I_{SENSE} \text{ threshold} = 1.33V$)



ELECTRICAL CHARACTERISTICS

ABSOLUTE MAXIMUM RATING		
Characteristics	Value	UNITS
Max V_{CC}	30	V
Operating junction temperature	- 40 to 150	°C
Storage temperature range	- 55 to 150	
Lead temperature (3 mm from case for 5 sec.)	260	

PARAMETER	TEST CONDITIONS	PARAMETERS			UNITS
		MIN.	TYP.	MAX.	
Voltage Regulation					
VSENSE threshold (note1)	$I_{CC} = 1 \text{ mA}$ (see figure 1, 2 & 5)	2.58	2.60	2.62	V
VSENSE threshold variation with temperature	$I_{CC} = 1 \text{ mA}$, -40°C to +85°C (see figure 5)	-	+/- 6	+/- 12	mV/°C
Output impedance of COMPV		-	1.2	-	KΩ
Voltage gain to COMPV	(see figure 7)	-	60	-	dB
Unity gain bandwidth	(see figure 7)	-	7	-	MHz
Phase margin in unity gain		-	70	-	Degrees
VSENSE Input current		-	-	0.6	μA
Transconductance from VSENSE to OPTO	@ $V_{CC} = 2.5V$ (see figure 1)	-	3.5	-	mA/mV
Max OPTO output sourcing current	@ $V_{CC} = 5V$ (see figure 1)	2	2.8	-	mA
Max OPTO output sourcing current	@ $V_{CC} = 2.5V$ (see figure 2)	1	1.3	-	mA
Current Limiting					
ISENSE threshold (note1)	$I_{CC} = 1 \text{ mA}$ (see figure 3 & 6)	1.31	1.33	1.35	V
ISENSE threshold variation with temperature	$I_{CC} = 1 \text{ mA}$, -40°C to +85°C (see figure 6)	-	+/- 3	+/- 6	mV
Output impedance of COMPI		-	400	-	KΩ
Voltage gain to COMPI	(see figure 8)	-	27	-	dB
Unity gain bandwidth to COMPI	(see figure 10)	-	500	-	KHz
Phase margin in unity gain		-	80	-	Degrees
Transconductance from ISENSE to OPTO	(see figure 8)	-	0.3	-	mA/mV
Max OPTO output sourcing current	@ $V(I_{SENSE}) = 1.33V$ (see figure 3)	3	4	-	mA

Note1: Tighter tolerances to 0.2% available upon request.

Note2: All values are @ 25°C unless otherwise specified.

Note3: Electrical parameters, although guaranteed, are not all 100% tested in production.

Fig 1: Transfer function of the voltage regulation

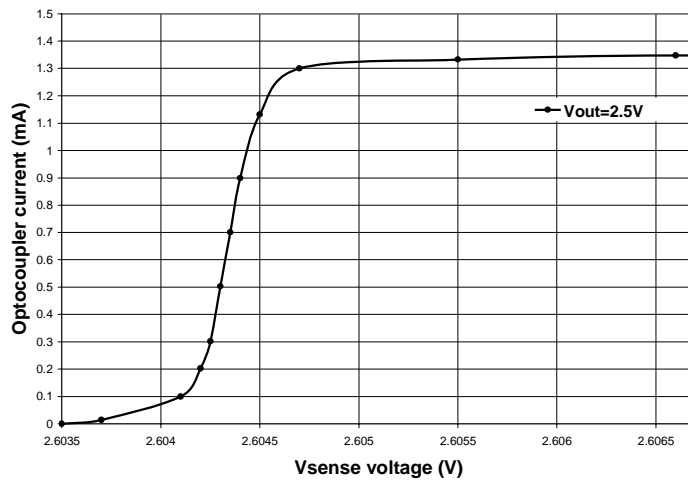


Fig 2: Transfer function of the voltage regulation

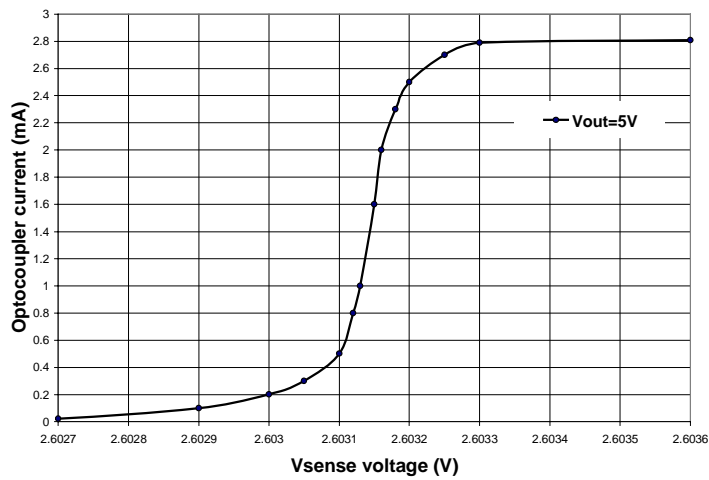


Fig 3: Transfer function of the current regulation

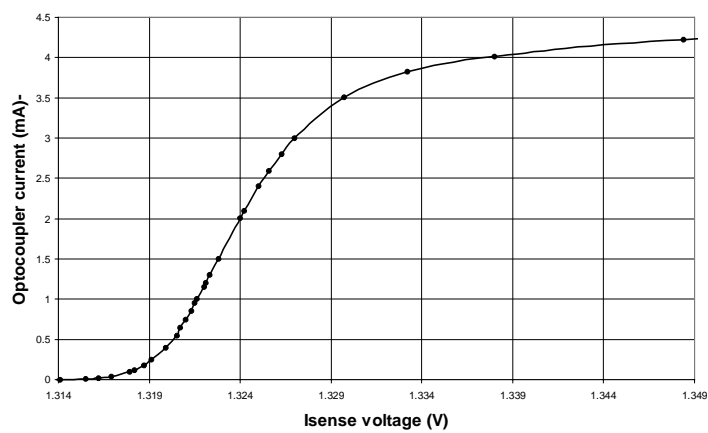


Fig 4: V/I profile using IPS25 with IPS10

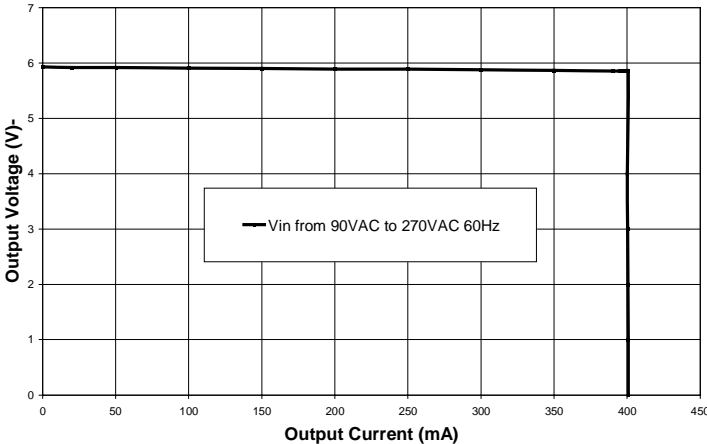


Fig 5: Voltage BGAP reference drift

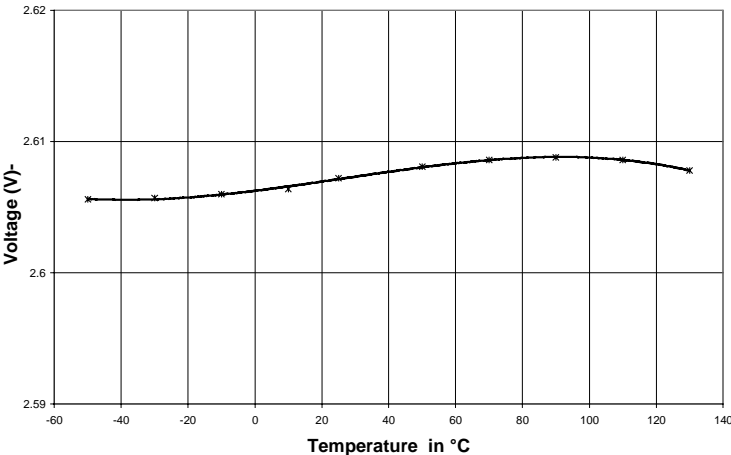
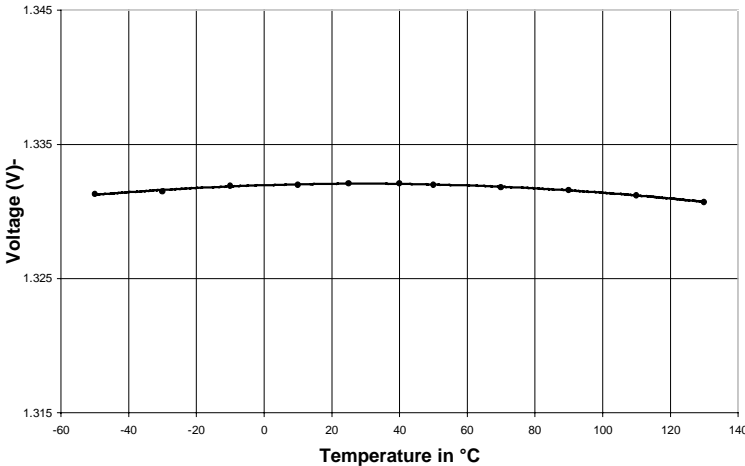


Fig 6: Current BGAP reference drift



OUTPUT RIPPLE Noise

when the **IPS25** is operating in conjunction with the IN-PLUG® **IPS10/IPS15** primary SMPS off-line driver IC.

(see application schematic on page 2)

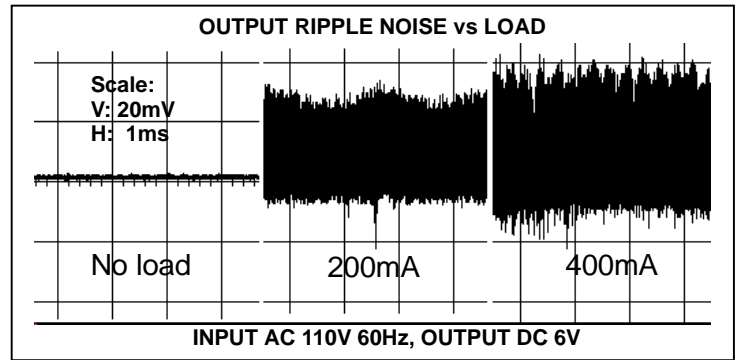


Fig 7: Open Loop gain of VSENSE amplifier with no compensation

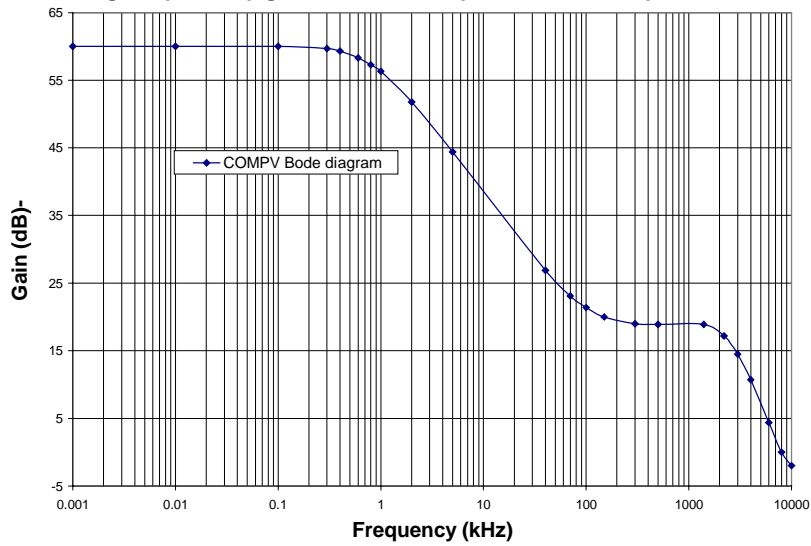


Fig 8: Open Loop Gain of ISENSE amplifier with no compensation

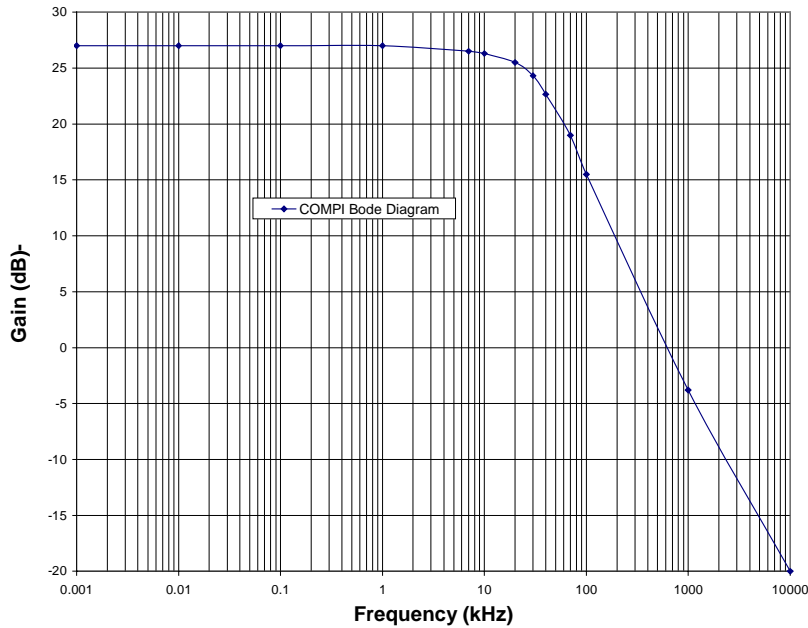


Fig 9: Frequency Roll-off of ISENSE amplifier with RC compensation

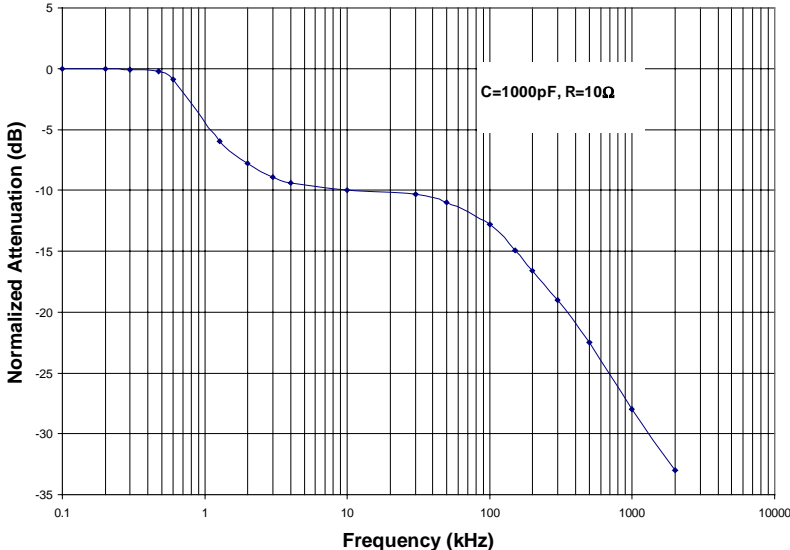
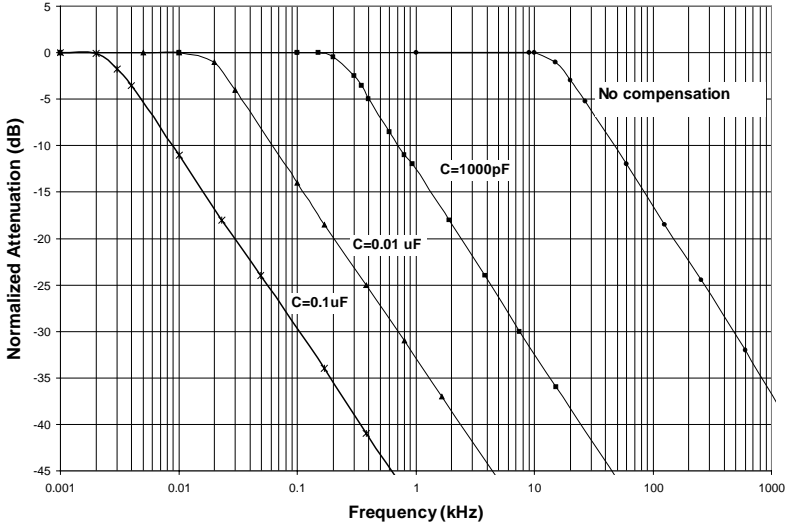
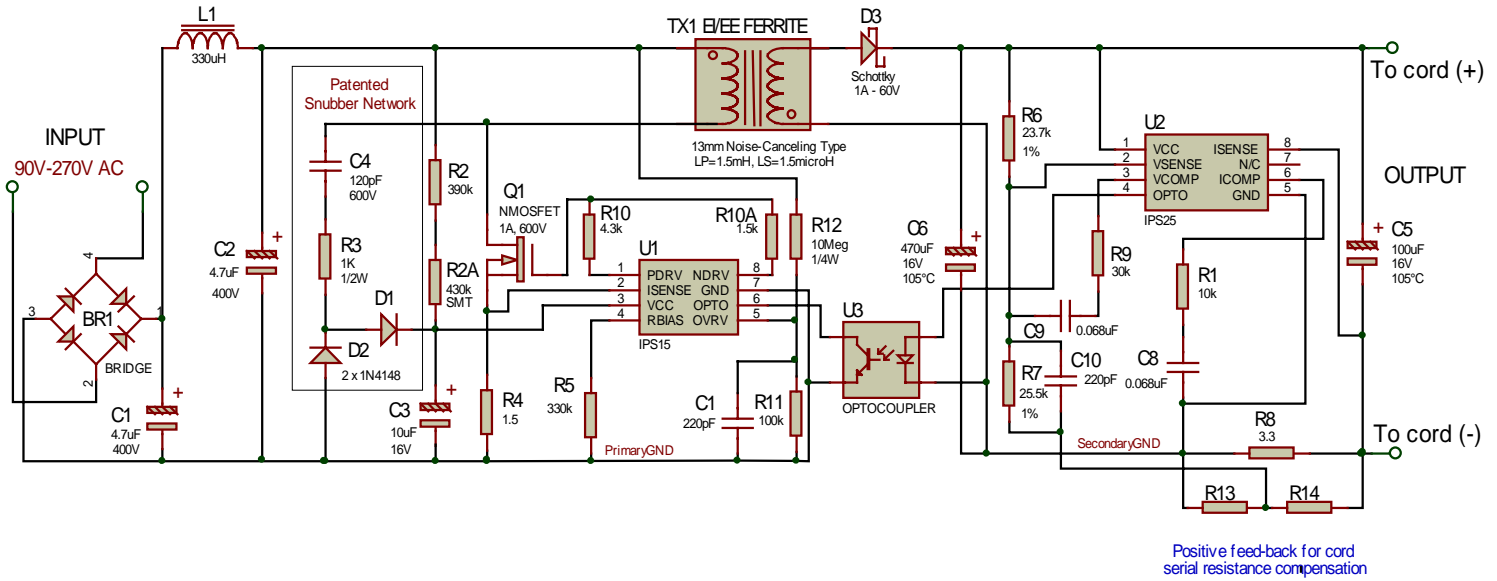


Fig 10: Frequency Roll-off of ISENSE amplifier with C compensation



OUTPUT CORD SERIAL RESISTANCE COMPENSATION



The voltage drop due to the output serial resistance of the cord can be compensated by adding a voltage positive feedback to the IPS25 input pin VSENSE, proportional to V_{drop} . This can be achieved either by splitting R8 (output current sensing resistor) in two or by inserting 2 serial resistors R13 and R14 in parallel to R8 and connect the positive feedback to the intermediate node.

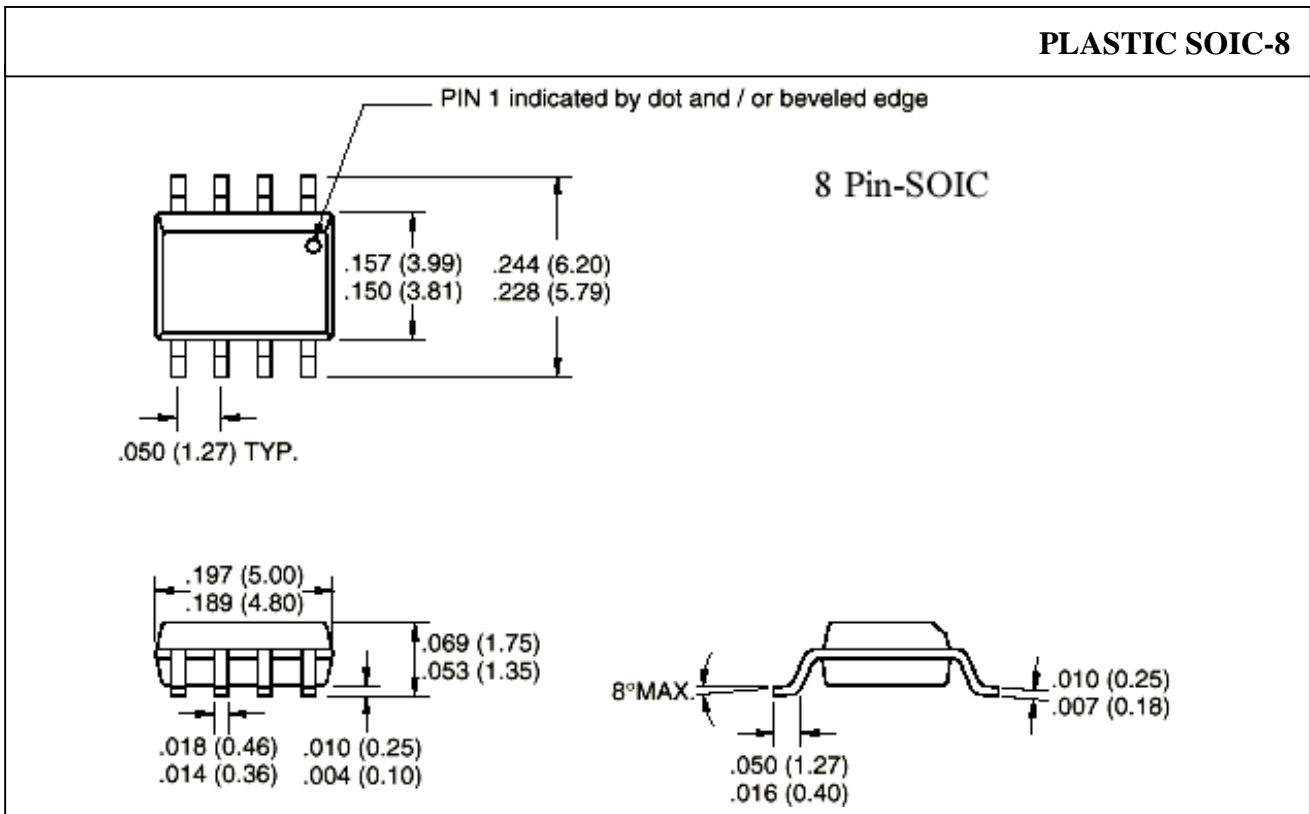
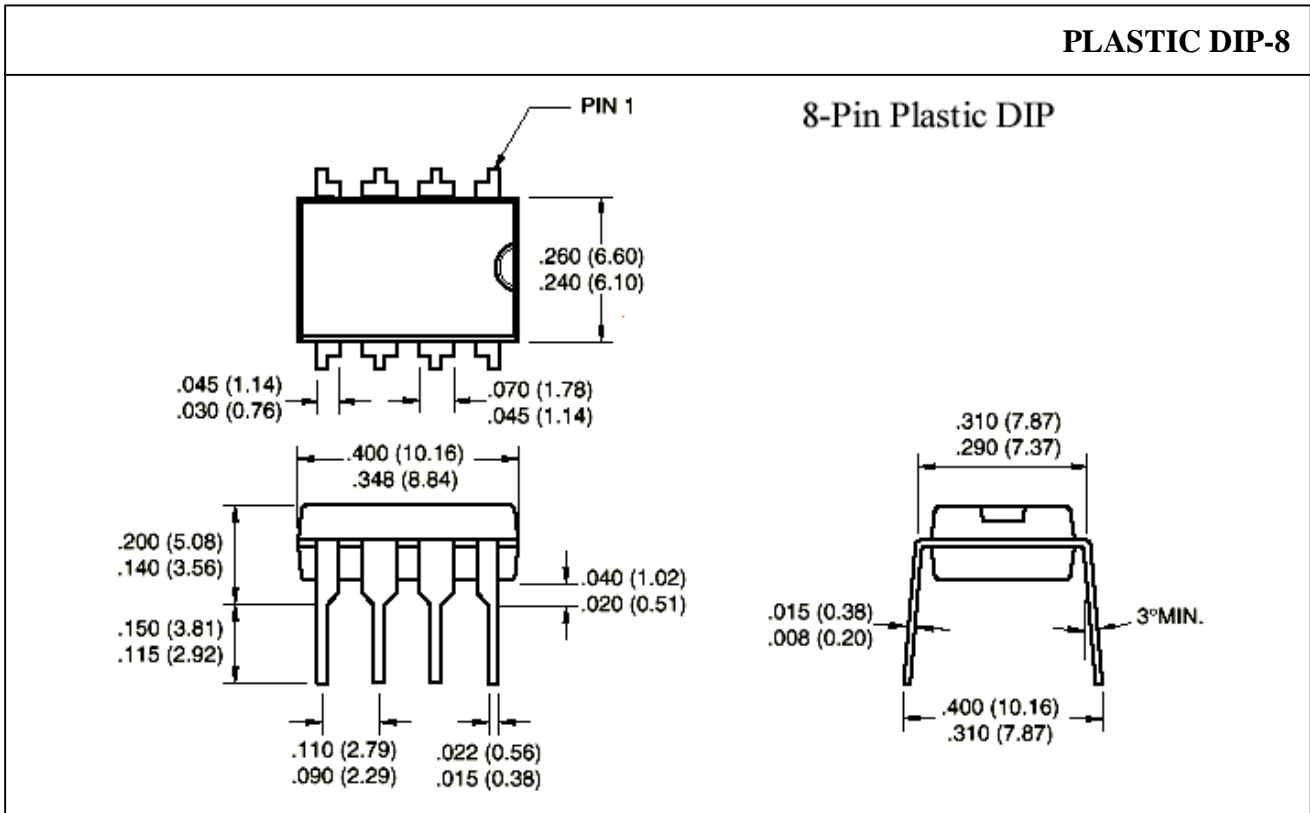
R13 and R14 should be calculated, based on the following conditions:

- $(R13 + R14) \geq 20 \times R8$ (but should remain low impedance in regards to R7),
- $V_{int} \sim V_{drop}$, where V_{int} is the intermediate voltage measured on the common node of R13 and R14, and V_{drop} is the voltage drop across the cord serial resistor.



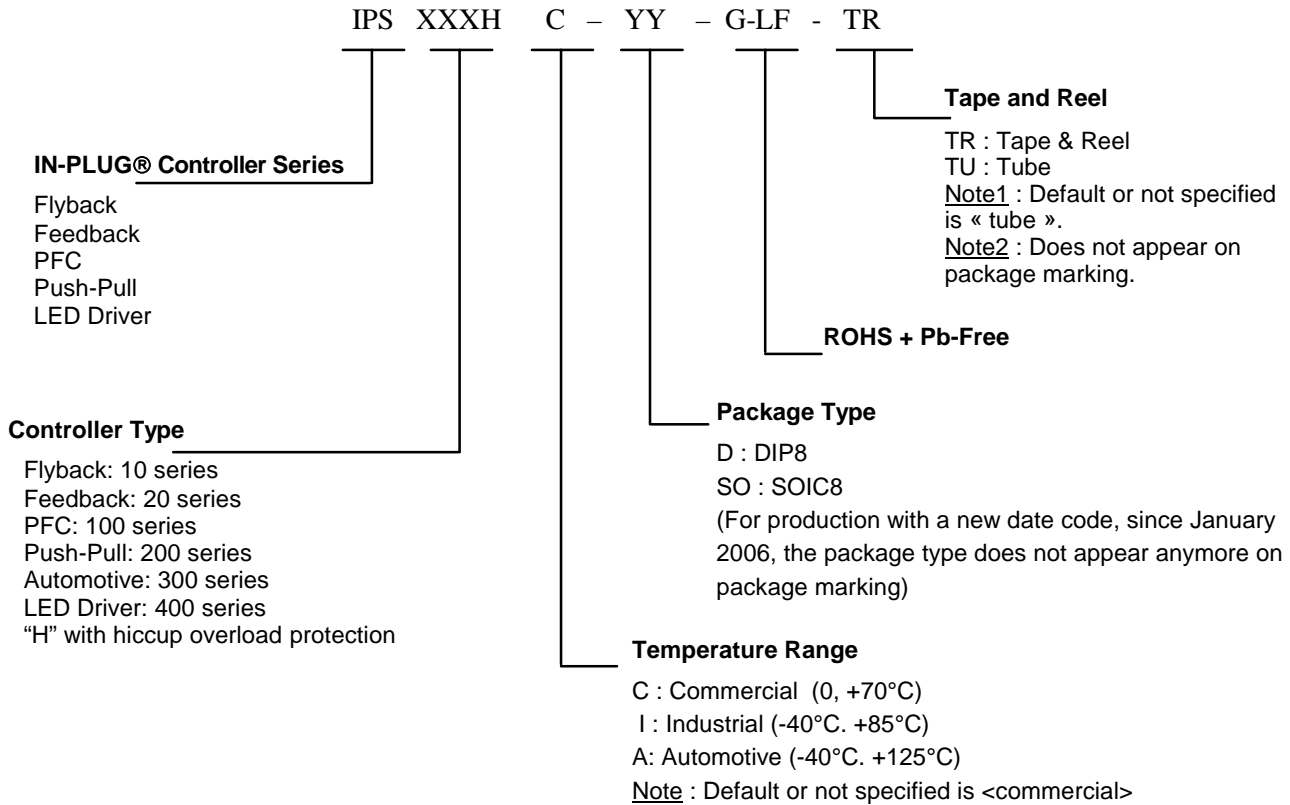
SMPS with output cord

PACKAGE DIMENSIONS



ORDERING INFORMATION

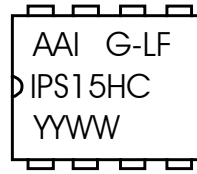
Part-Number



Example of Marking



Non-Green Package



Green ROHS + Pb-Free Package

(Note : For production with a new date code, since January 2006, the package type does not appear anymore on package marking)

This ordering information is for commercial and industrial standard IN-PLUG® controllers ONLY. For custom controllers or for military temperature range, call AAI's sales representative.

IN-PLUG® IPS25 Datasheet - Rev.9 - Voltage / Current Feedback Controller

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- AAI's modified snubber network is patented under the US Patent # 6,233,165

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ASIC Advantage INC.

1290-B Reamwood Ave, Sunnyvale California 94089, USA

Tel: (1) 408-541-8686 Fax: (1) 408-541-8675

Websites: <http://www.in-plug.com> - <http://www.asicadvantage.com>